

EECE 301  
Signals & Systems  
Prof. Mark Fowler

**Note Set #30**

- D-T Systems: IIR Filters

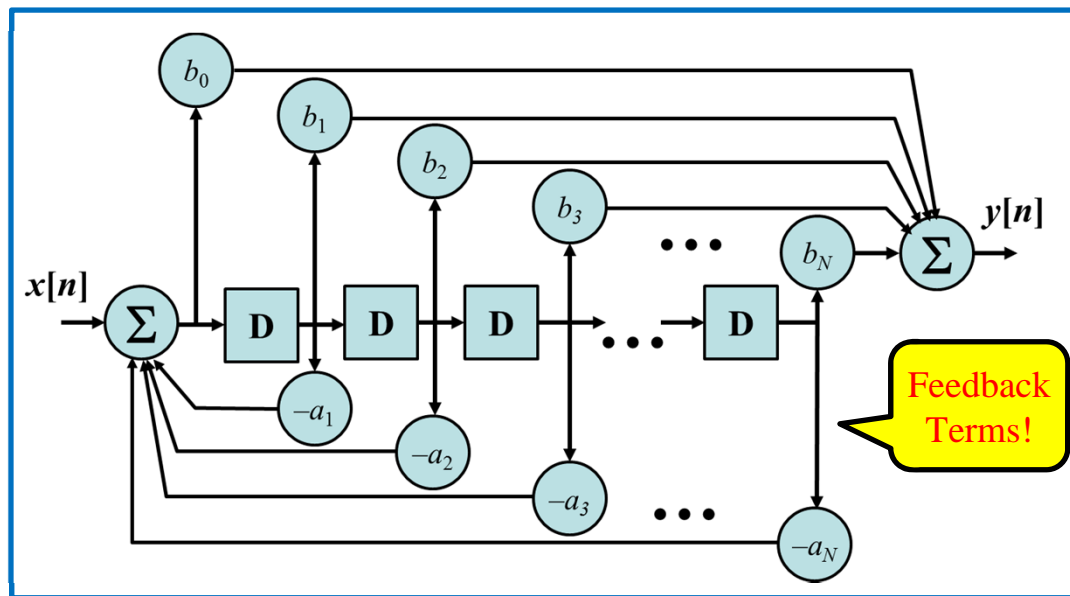
# IIR Filters (Recursive Filters)

IIR (Recursive) filters have two issues that constrain their use...

1. They do not have linear phase (*can* get approx. linear phase w/ special designs)
  - linear phase is more crucial in certain areas...like digital comm & radar (which involve pulses) or filtering images (which involves edges).
2. May not be inherently stable (feedback gives poles other than at origin)
  - This can be a serious issue when implementing IIR filters

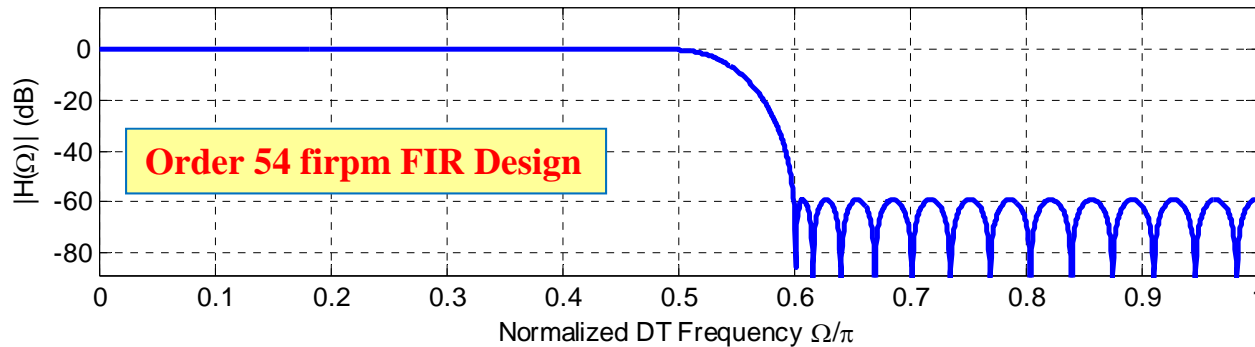
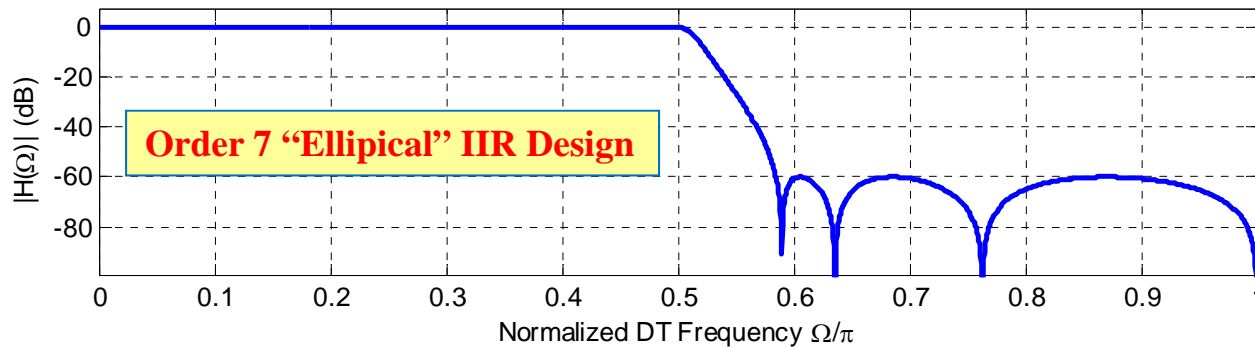
IIR Filters have one main advantage over FIR filters:

- can get good magn. resp. w/o high computational complexity



**You can usually get quite good filters even with fairly low orders (like 10 or so).**

# Complexity Comparison: IIR vs FIR



This IIR requires:

- 15 multiplies
- 13 additions

This FIR requires:

- 55 multiplies
- 54 additions

Almost 4x as much computation for the FIR filter!

# MATLAB-Based IIR Design

MATLAB has several easy commands for IIR design, including:

- butter, cheby1, cheby2, ellip

## Chebyshev IIR

```
[b,a]=cheby2(7,60,0.7);
```

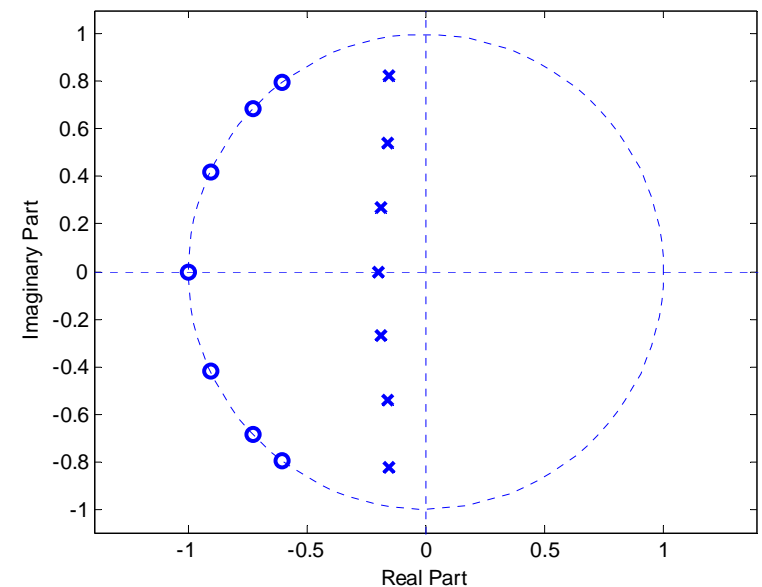
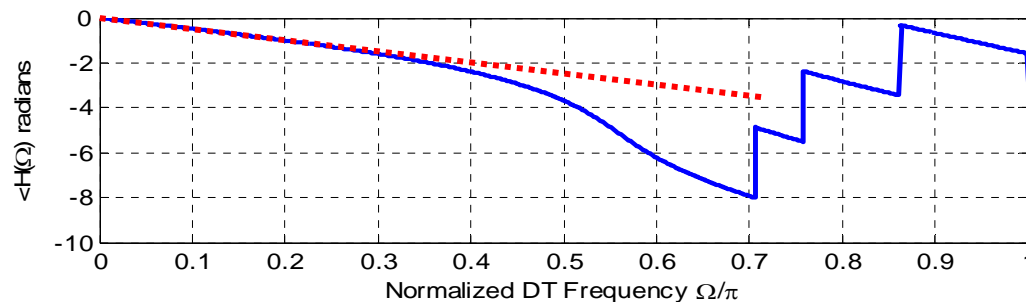
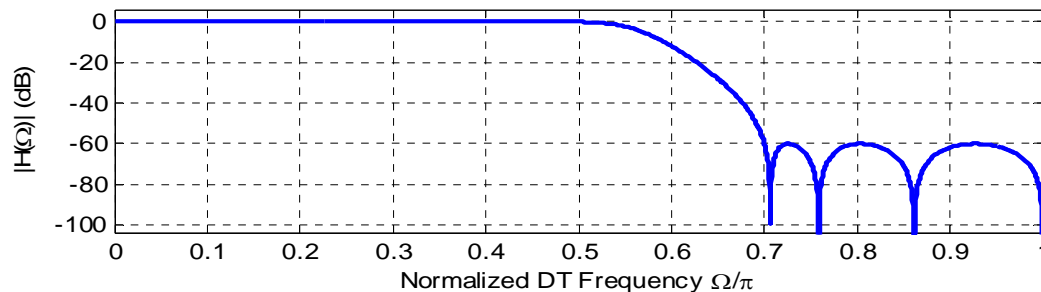
Sets Order

Sets Stopband Height

Sets Stopband Edge

**b = [0.0692 0.3789 0.9728 1.5028 1.5028 0.9728 0.3789 0.0692]**

**a = [1.0000 1.2028 1.6599 1.0991 0.6240 0.2098 0.0473 0.0048]**



# Elliptical IIR

```
[b,a] = ellip(7,0.1,60,0.5);
```

Sets Order

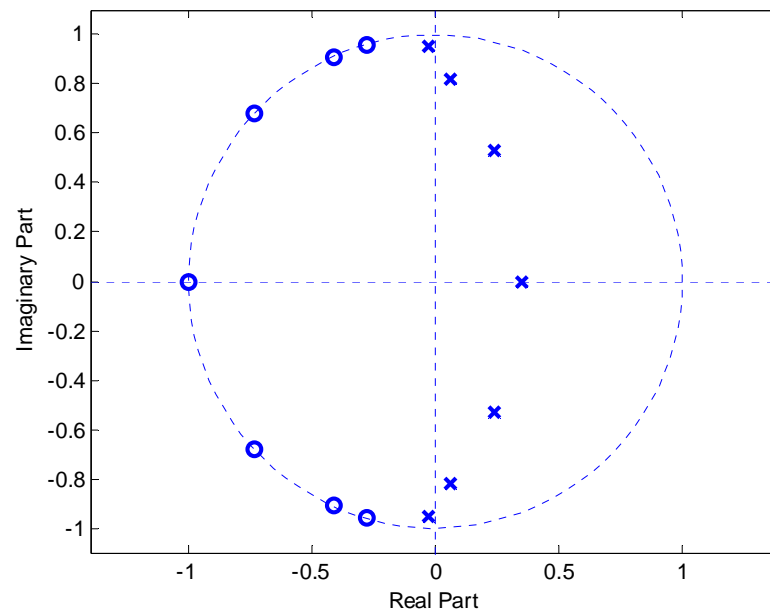
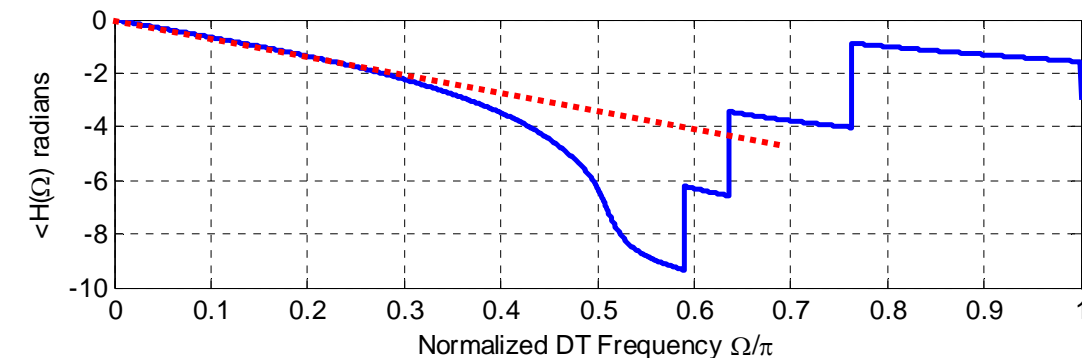
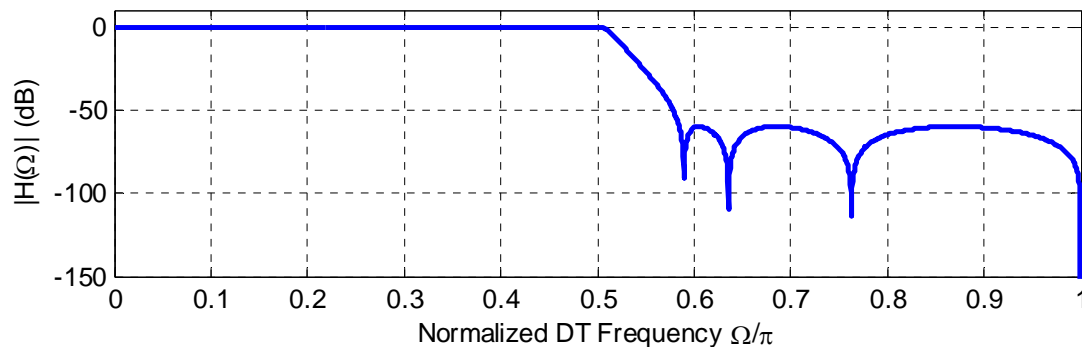
Sets Passband Ripple

Sets Stopband Height

Sets Passband Edge

**b = [0.0338 0.1302 0.2821 0.4013 0.4013 0.2821 0.1302 0.0338]**

**a = [1.0000 -0.8994 2.1386 -1.5364 1.4793 -0.7327 0.3178 -0.0725]**



# Butterworth IIR

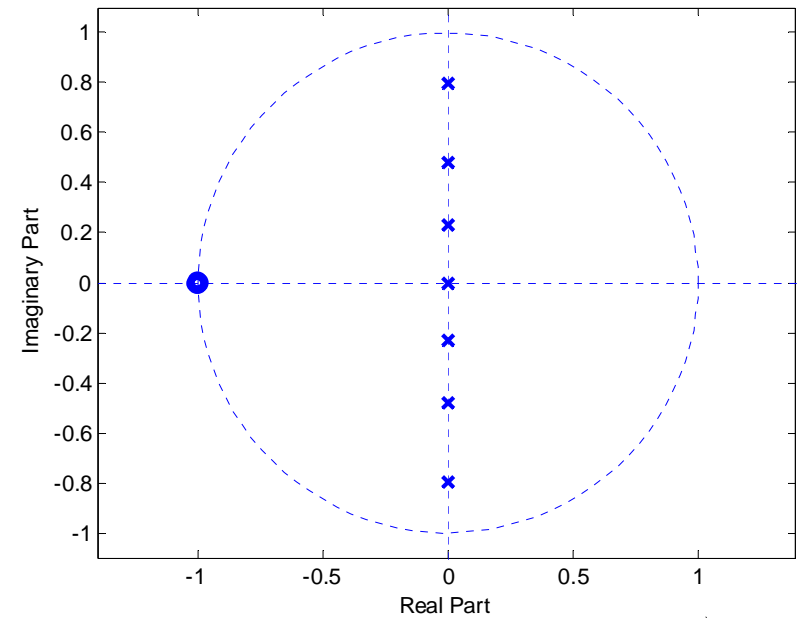
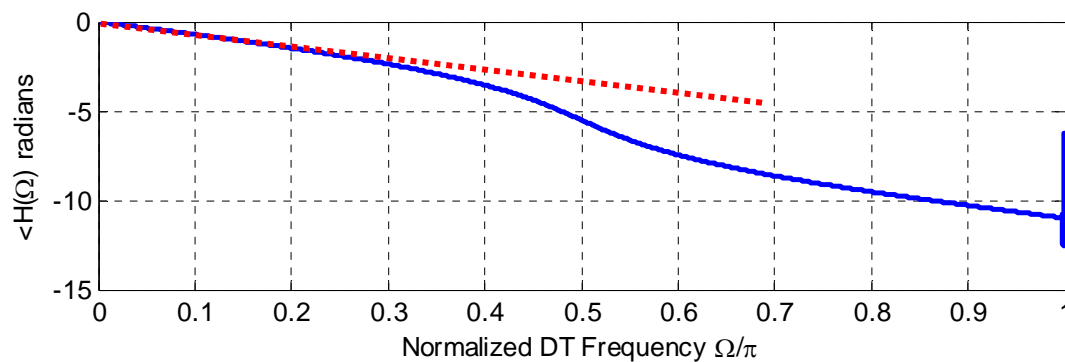
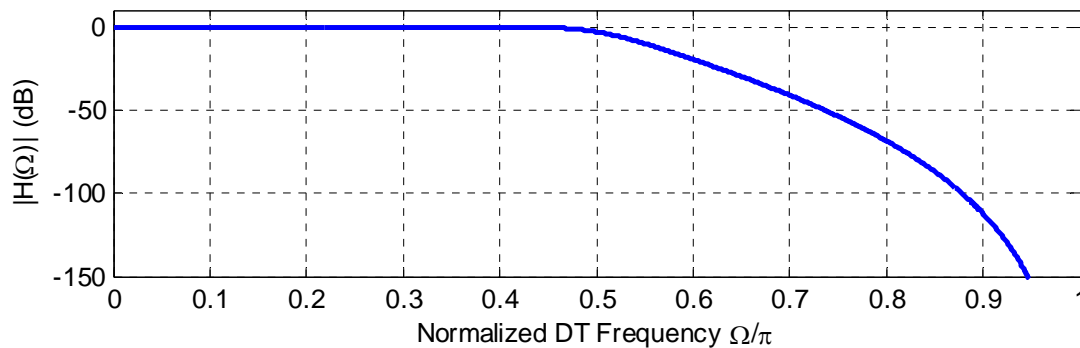
```
[b,a] = butter(7,0.5);
```

Sets Order

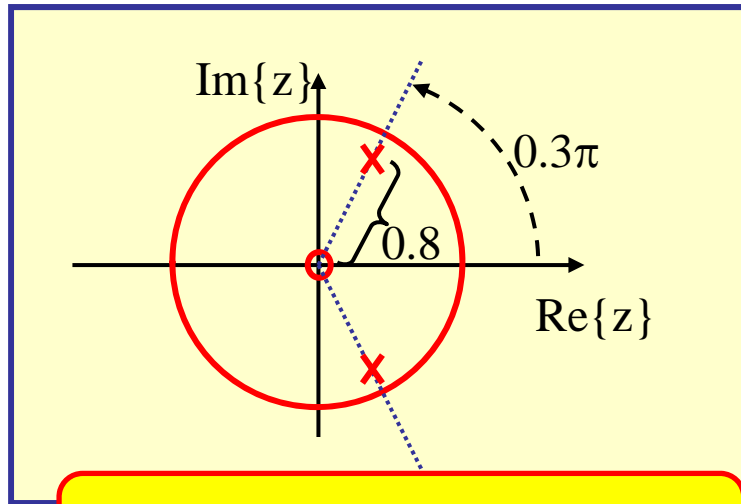
Sets Passband Edge

**b = [0.0166 0.1160 0.3479 0.5798 0.5798 0.3479 0.1160 0.0166]**

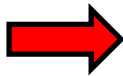
**a = [1.0000 -0.0000 0.9200 -0.0000 0.1927 -0.0000 0.0077 -0.0000]**



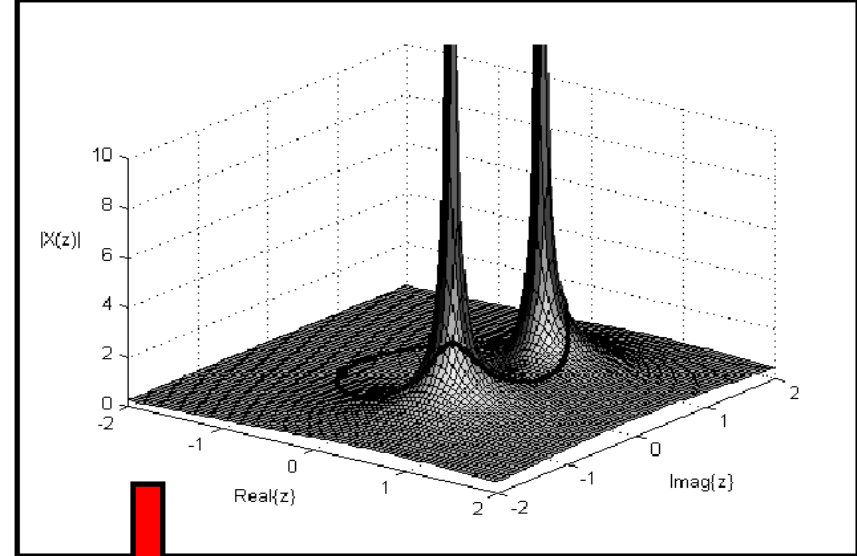
# Visualizing IIR Freq Resp.



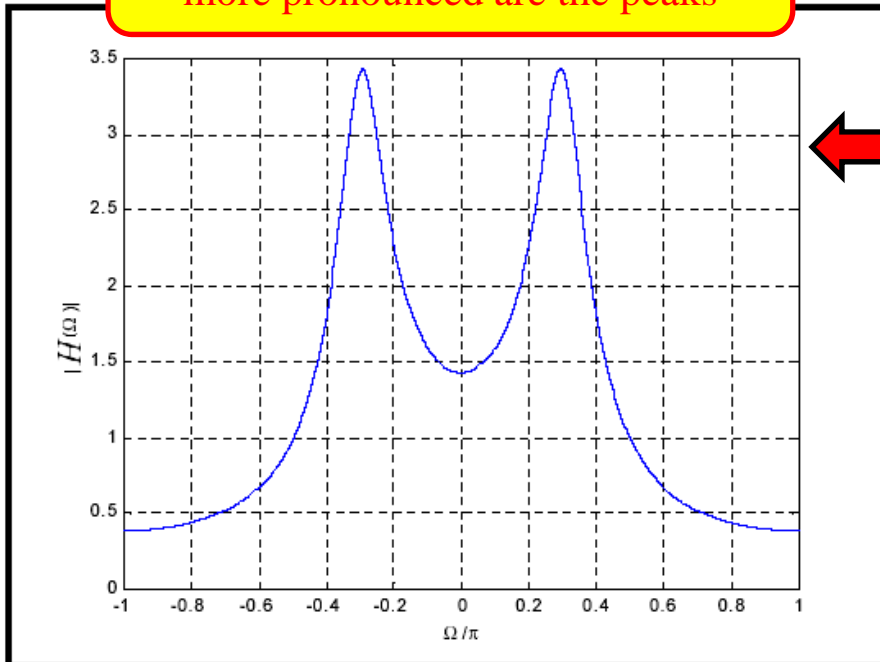
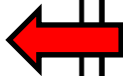
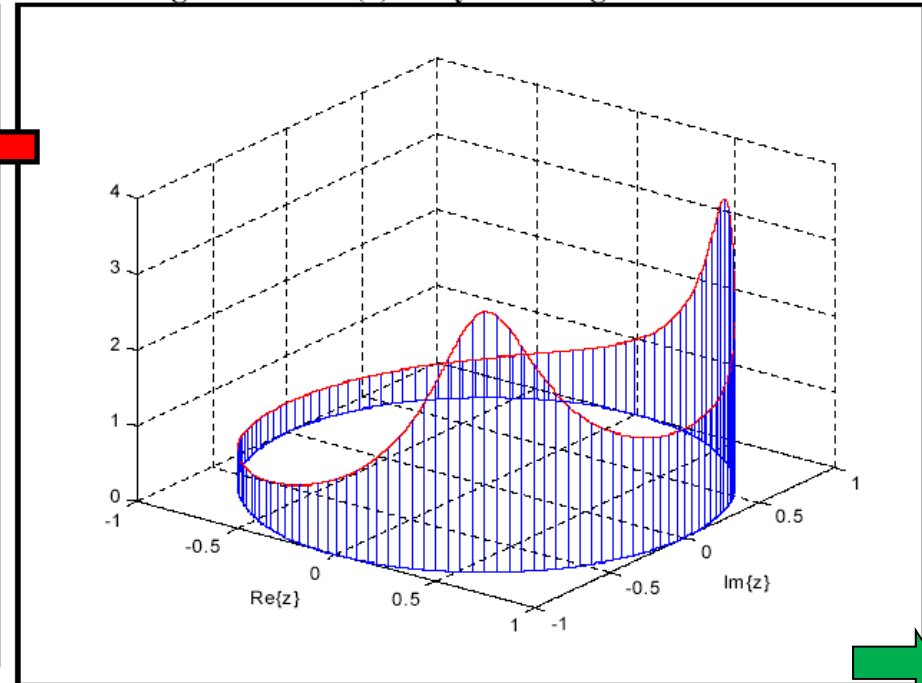
Closer the poles are to the UC the more pronounced are the peaks



Surface Plot of Magnitude of  $H(z)$ ; Shows Values on Unit Circle

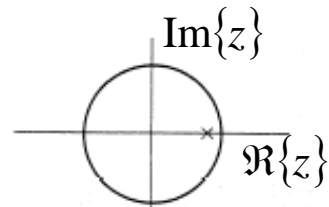


Plot of Magnitude of  $H(z)$  Only Showing Values on Unit Circle

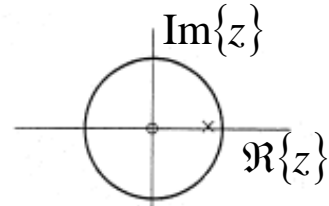
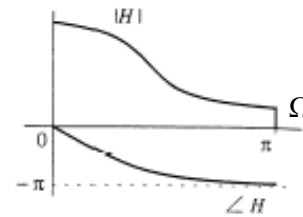


# Effect of Poles & Zeros on Frequency Response of DT filters

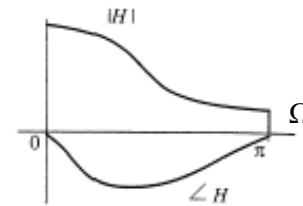
Note: Including a pole or zero at the origin ...



(a)

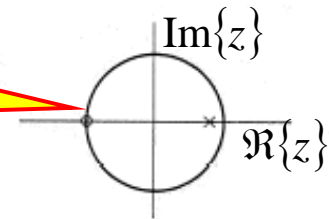


(b)

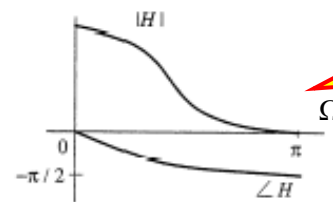


...doesn't change the magnitude but does change the phase

Placing a zero on UC...

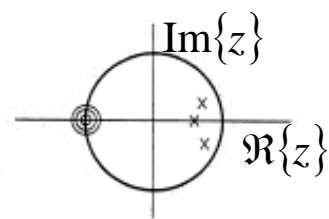


(c)

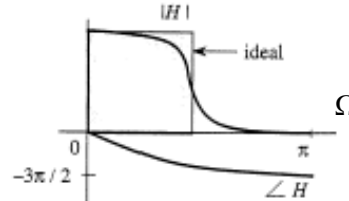


...makes  $|H| = 0$  at angle where zero is placed

Placing more zeros/poles...



(d)



... gives sharper transitions.

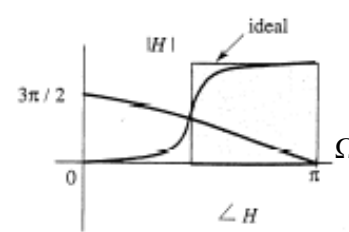
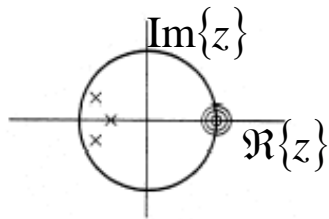


Figure from B.P. Lathi, Signal Processing and Linear Systems



### Effect of Zero at Origin

$H_1(z)$  has an extra zero at the origin:

$$H_1(z) = zH(z)$$

$$H_1(\Omega) = e^{j\Omega} H(\Omega)$$

$$\begin{aligned} |H_1(\Omega)| &= |e^{j\Omega} H(\Omega)| \\ &= |e^{j\Omega}| |H(\Omega)| \\ &= |H(\Omega)| \end{aligned}$$

No Effect on  
Magnitude

$$\begin{aligned} \angle H_1(\Omega) &= \angle e^{j\Omega} H(\Omega) \\ &= \angle H(\Omega) + \Omega \end{aligned}$$

Effect on Phase

### Effect of Pole at Origin

$H_1(z)$  has an extra pole at the origin:

$$H_1(z) = z^{-1} H(z)$$

$$H_1(\Omega) = e^{-j\Omega} H(\Omega)$$

$$\begin{aligned} |H_1(\Omega)| &= |e^{-j\Omega} H(\Omega)| \\ &= |e^{-j\Omega}| |H(\Omega)| \\ &= |H(\Omega)| \end{aligned}$$

$$\begin{aligned} \angle H_1(\Omega) &= \angle e^{-j\Omega} H(\Omega) \\ &= \angle H(\Omega) - \Omega \end{aligned}$$

# Effect of Feedback

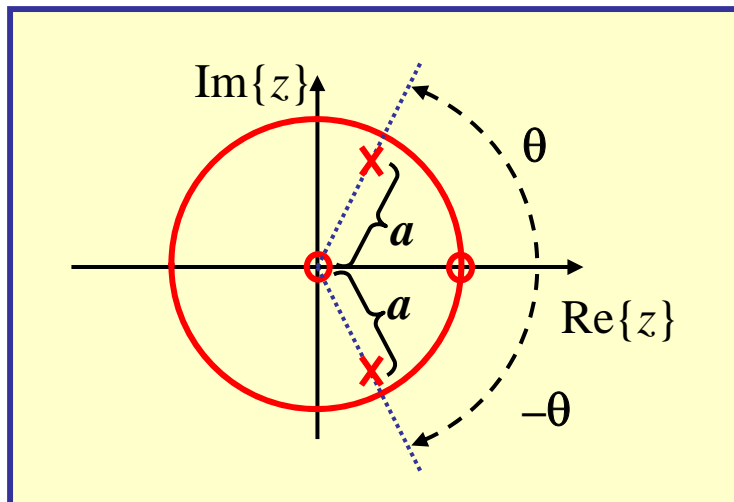
We know two important things:

- Feedback is what can give us poles other than at the origin
- Any pole outside the UC causes the system to be unstable

**Thus.... It is feedback that can cause a system to be stable.**

$$y[n] - 2a \cos(\theta) y[n-1] + a^2 y[n-2] = x[n] - x[n-1] \quad \text{For } a > 0 \text{ and } 0 \leq \theta \leq \pi$$

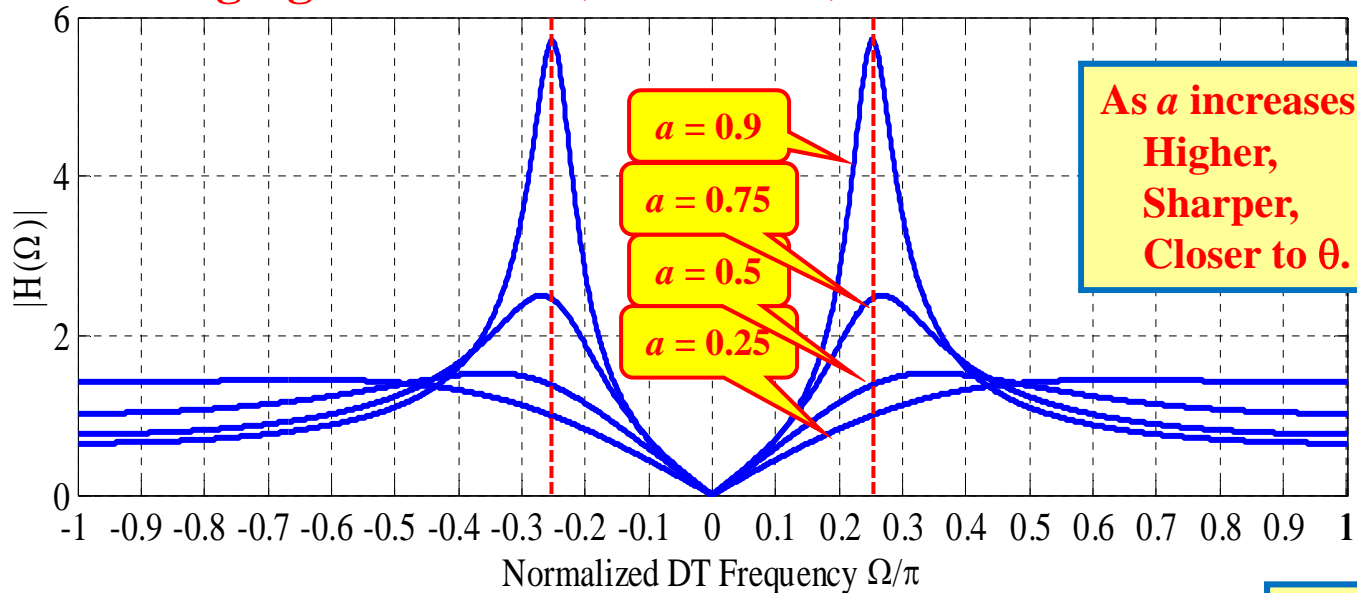
$$H(z) = \frac{1 - z^{-1}}{1 - 2a \cos(\theta) z^{-1} + a^2 z^{-2}} = \frac{z(z-1)}{z^2 - 2a \cos(\theta)z + a^2}$$



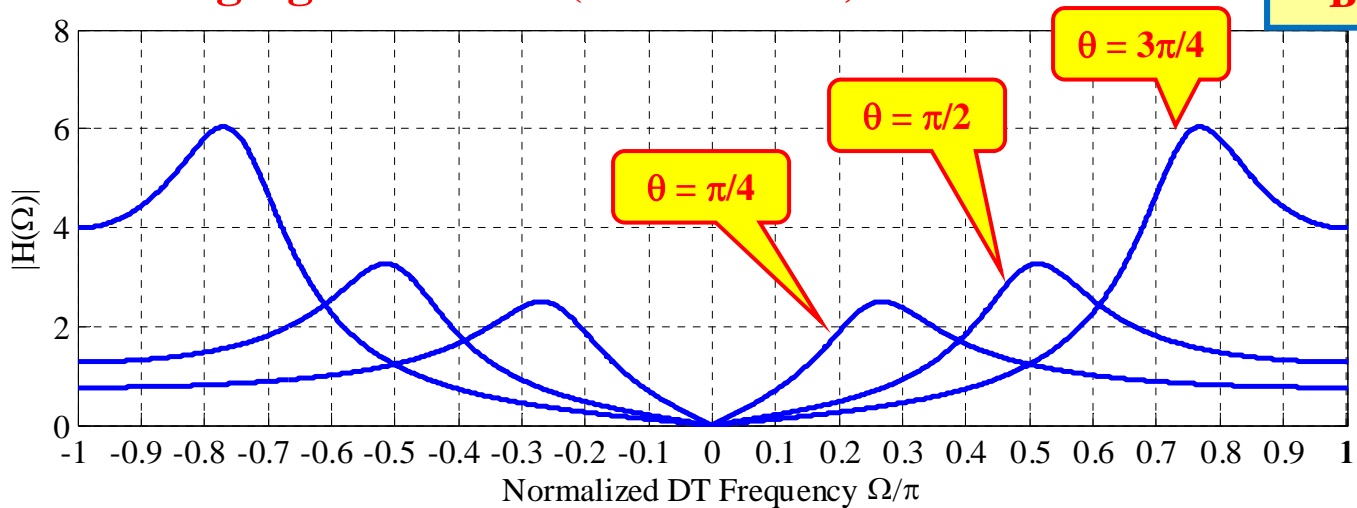
Has roots (poles!) at:  
 $z = ae^{\pm j\theta}$

**So... if  $a$  is increased to be 1 or larger then the system is unstable!**

## Effect of Changing Value of $a$ (with $\theta = \pi/4$ ):



## Effect of Changing Value of $\theta$ (with $a = 0.75$ ):



WHY???

