

EECE 301  
Signals & Systems  
Prof. Mark Fowler

**Note Set #9**

- C-T Signals: FS Spectrum

# Trig Form “Spectrum”... Is “Single Sided”

Best for “thinking about real-world ideas”

Trig Form: Amplitude & Phase

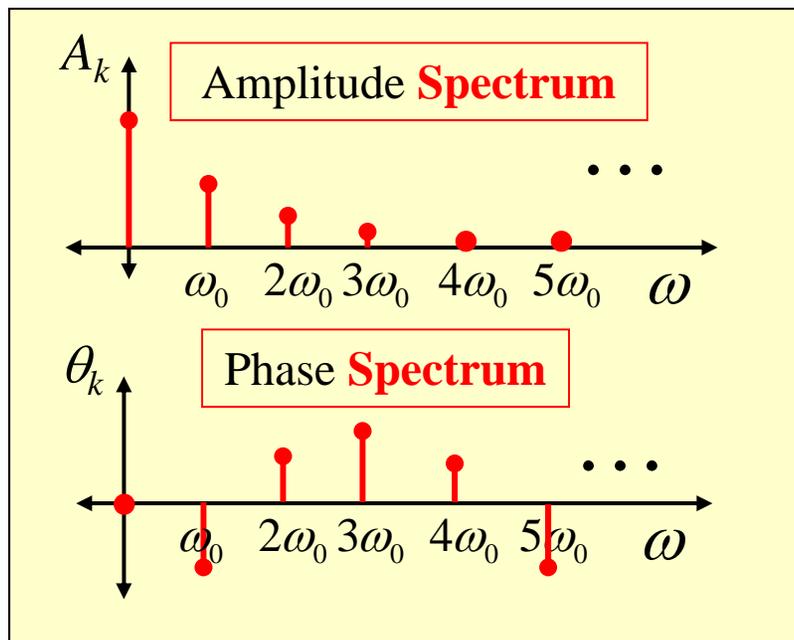
$$x(t) = A_0 + \sum_{k=1}^{\infty} A_k \cos(k\omega_0 t + \theta_k)$$

Need  $A_k$  and  $\theta_k$   
for  $k = 0, 1, 2, \dots$

$A_k$  = Amplitude  
 $\theta_k$  = Phase

So... to describe a signal via FS we specify:  
“Amplitude & Phase @ Each Frequency”

A good way to “see” the FS coefficients is by plotting them vs. frequency:



For this form of FS:

- Do not need negative freqs  
→ “Single Sided” Spectrum

# Exp Form “Spectrum”... Is “Double Sided”

Best for “doing math” ( $c_k$  are like phasors!!)

**Exponential Form**

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

Need  $c_k$  (complex!)  
for  $k = \dots -2, -1, 0, 1, 2 \dots$

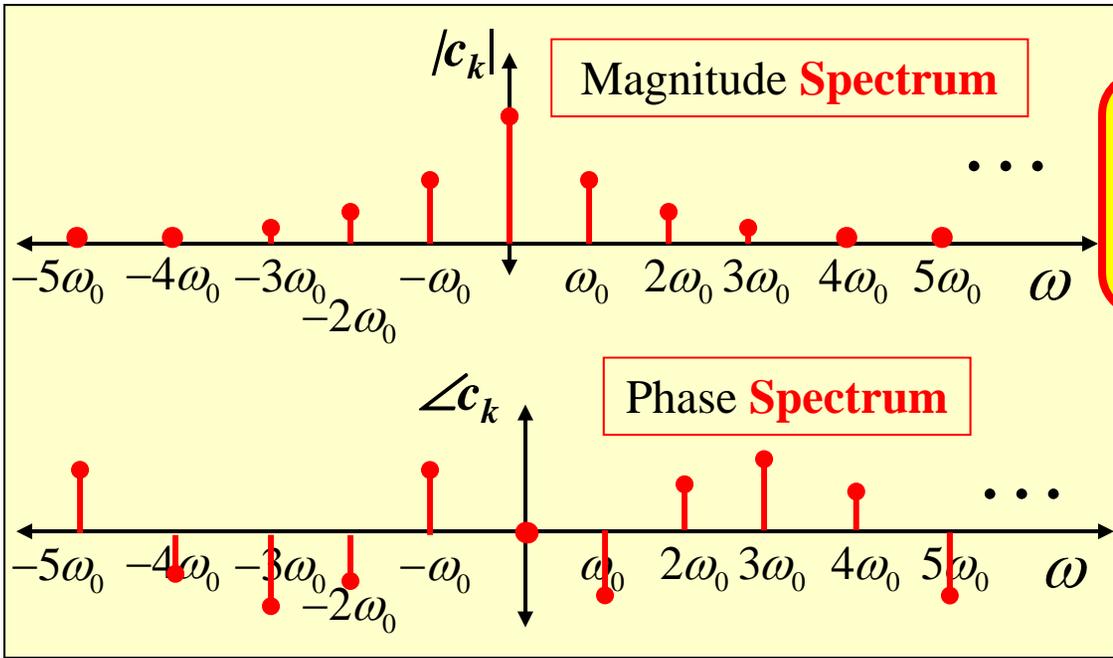
$|c_k|$  = Magnitude  
 $\angle c_k$  = Phase

$$c_k e^{jk\omega_0 t} = \left[ |c_k| e^{j\angle c_k} \right] e^{jk\omega_0 t}$$

$$= |c_k| e^{j(k\omega_0 t + \angle c_k)}$$

So... to describe a signal via FS we specify:  
“Magnitude & Phase @ Each Frequency”

For this form of FS:  
• Do need negative freqs  
→ “Double Sided” Spectrum



# Spectrum Characteristics

## Trig Form: Amplitude & Phase

$$x(t) = A_0 + \sum_{k=1}^{\infty} A_k \cos(k\omega_0 t + \theta_k)$$

For Trig Form of FS Spectrum:

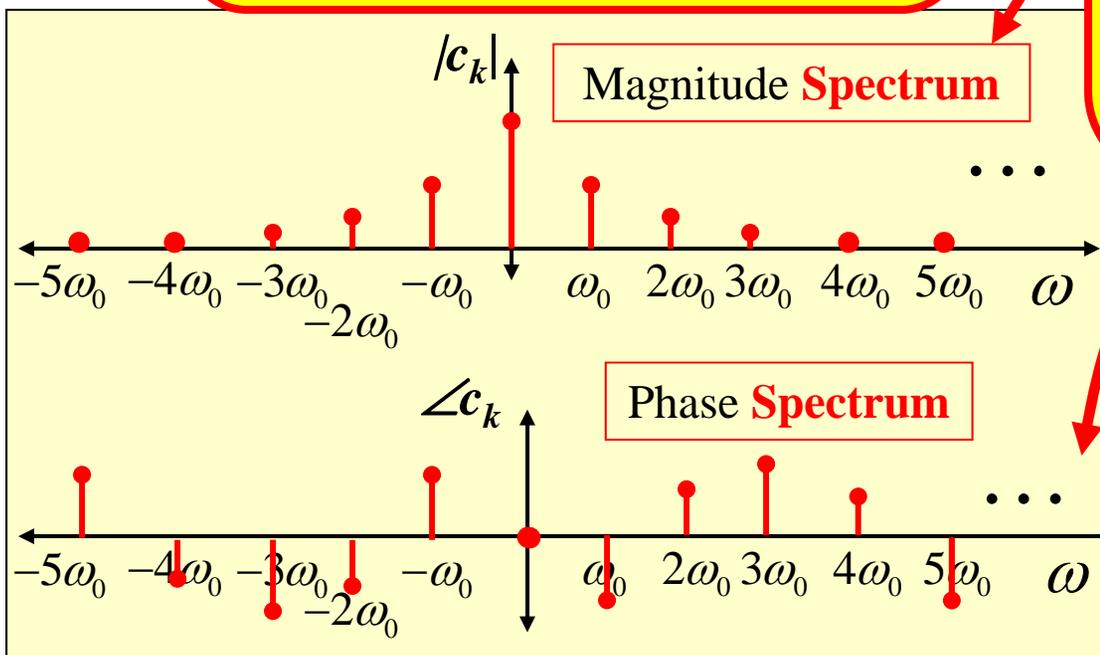
- **“Single Sided” Spectrum**
- $A_k \geq 0$  for  $k > 0$ 
  - $A_0$ : positive or negative
- $\theta_k$  is in **radians**  $\theta_0 = 0$

## Exponential Form

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t}$$

For Exp Form of FS Spectrum:

- **“Double Sided” Spectrum**
- $|c_k| \geq 0$  for all  $k$ 
  - **Even Symmetry for Magn.**
- $\angle c_k$  is in **radians**
- $\angle c_0 = 0$  or  $\pm\pi$
- $\angle c_k = -\angle c_{-k}$ 
  - **Odd Symmetry for Phase**



$$\left. \begin{aligned} c_k &= \frac{1}{2} A_k e^{j\theta_k} \\ c_{-k} &= \frac{1}{2} A_k e^{-j\theta_k} \end{aligned} \right\} k = 1, 2, 3, \dots$$

# Parseval's Theorem

We saw earlier how to compute the average power of a periodic signal if we are given its time-domain model:

$$P = \frac{1}{T} \int_{t_0}^{T+t_0} x^2(t) dt$$

**Q: Can we compute the average power from the frequency domain model**

**A: Parseval's Theorem says... Yes!**

$$\{c_k\}, \quad k = 0, \pm 1, \pm 2, \dots$$

Parseval's theorem says that the avg. power can be computed this way:

$$P = \sum_{k=-\infty}^{\infty} |c_k|^2$$



$$\frac{1}{T} \int_{t_0}^{t_0+T} x^2(t) dt = \sum_{k=-\infty}^{\infty} |c_k|^2$$

$c_k$  are the Exp. Form FS coefficients

Left side is clearly finite for real-world signals...

**Thus, the  $|c_k|$  must decay fast enough as  $k \rightarrow \pm\infty$**

**Tells us something about how the magnitude spectrum should look!**

# Interpreting Parseval's Theorem

$$\underbrace{\frac{1}{T} \int_{t_0}^{t_0+T} x^2(t) dt}_{\text{“sum” of squares in time-domain model}} = \underbrace{\sum_{k=-\infty}^{\infty} |c_k|^2}_{\text{“sum” of squares in freq.-domain model}}$$

“sum” of squares in time-domain model

“sum” of squares in freq.-domain model

$x^2(t)$  = power at time  $t$  (includes effects of all frequencies)

We can find the power in the time domain by “adding up” all the “powers at each time”

$|c_k|^2$  = power at frequency  $k\omega_0$  (includes effects of all times)

We can find the power in the frequency domain by adding up all the “powers at each frequency”

# One Use for Parseval's Theorem

When numerically computing the FS approximation... PT allows you to compute the power of the error term:

First find Avg Power:

$$P = \frac{1}{T} \int_{t_0}^{T+t_0} x^2(t) dt$$

Do analytically or numerically

Then find power of approximate using PT:

$$P_{approx} = \sum_{k=-K}^K |c_k|^2 = |c_0|^2 + 2 \sum_{k=1}^K |c_k|^2$$

Then find power of error as  $P_{error} = P - P_{approx}$

It is easy to show that

$$P_{error} = 2 \sum_{k=K+1}^{\infty} |c_k|^2$$

Since the  $|c_k|$  decay as  $k \rightarrow \infty$  this shows that we can make  $P_{error}$  as small as we want by making  $K$  big enough!

